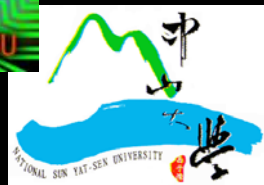
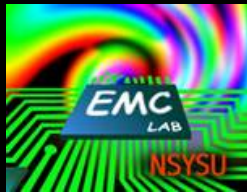


A New Electromagnetic Bandgap Power Plane with super-broadband suppression of Ground Bounce Noise in High Speed Circuits



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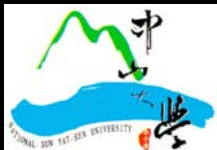
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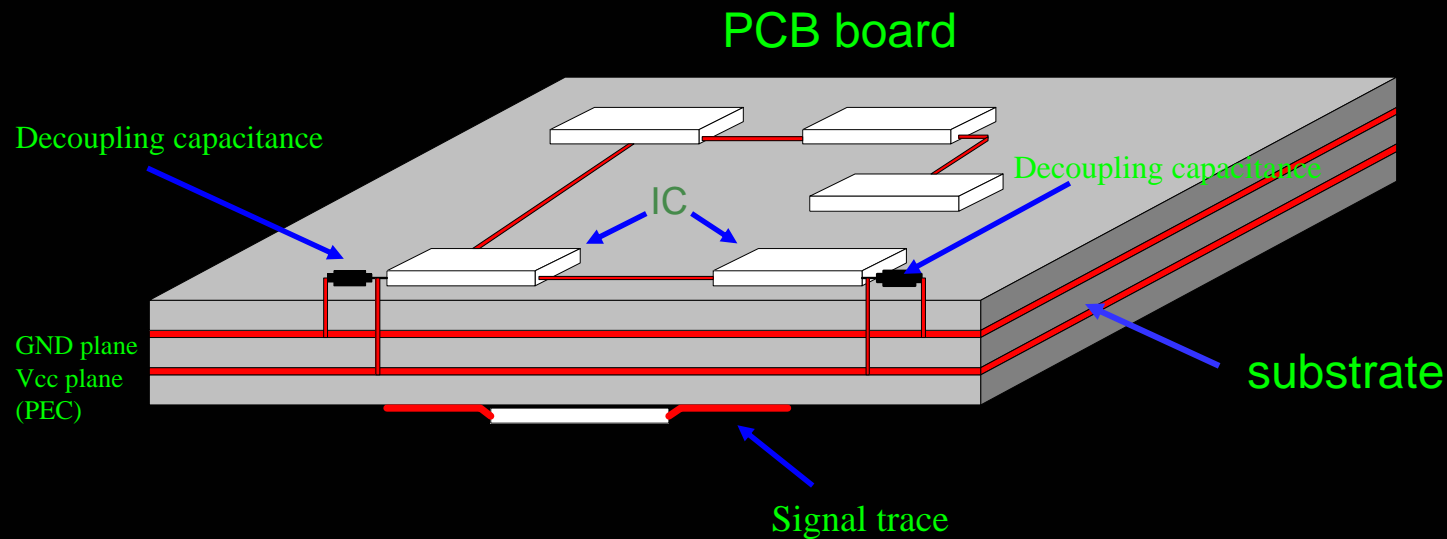
Outline

- Introduction
- Design Concept of L- bridged EBG Power Planes
- The Simulation and Measurement Results
- Stopband Model for the EBG Power Planes
- Signal integrity for L-bridge EBG power plane
- EMI for L-bridge EBG power plane
- Conclusion

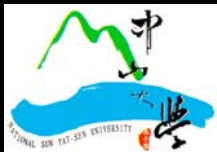


Introduction

- Ground bounce noise (GBN) on the power/ground planes is becoming one of the major concerns for the high-speed digital computer systems.
- Adding decoupling capacitors to suppress the GBN[7] is not effective at frequencies higher than 600MHz due to their finite lead inductance.

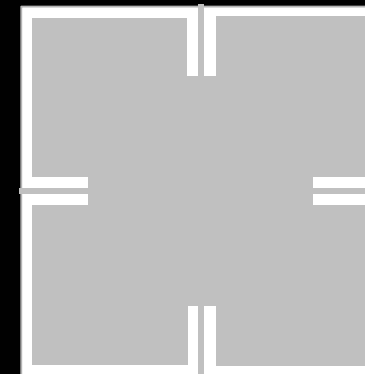
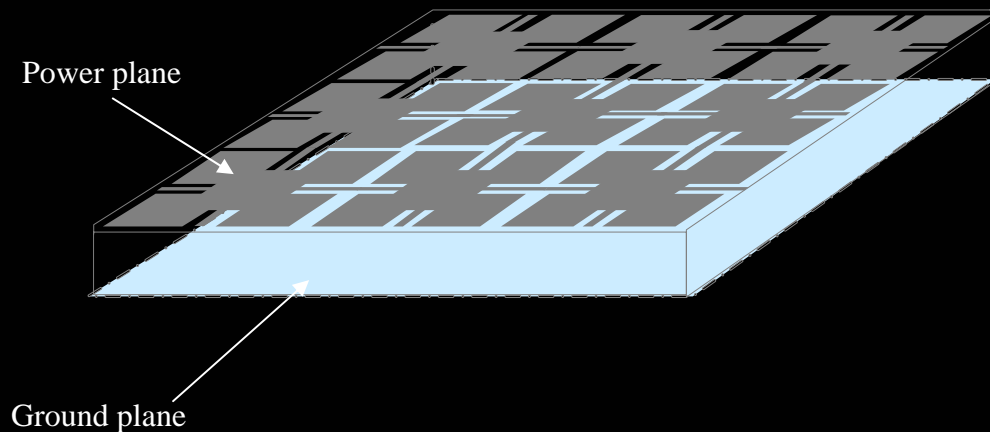


[7] Tzong-Lin Wu, Yen-Hui Lin, Jiuun-Nan Hwang and Jig-Jong Lin, "The effect of test system impedance on measurements of ground bounce in printed circuit boards" *IEEE Transactions on Electromagnetic Compatibility*, vol. 43 ,No. 4 ,pp. 600 – 607, Nov. 2001



Introduction

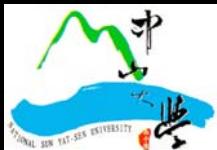
- A new idea for eliminating the GBN is proposed by designing electromagnetic bandgap (EBG) structure on the power or ground plane.
- The first EBG power/ground plane design[5], [8] was demonstrated with 1.7GHz stop-band bandwidth centered at 3.77GHz.



Unit cell

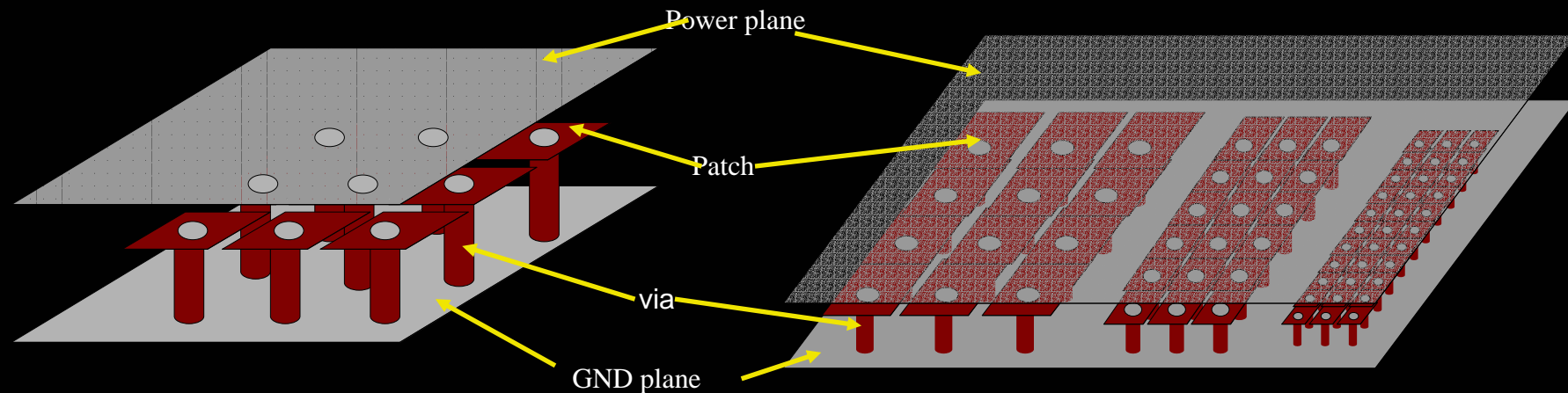
[5] F. R. Yang, K. P. Ma, Y. Q. and T. Itoh, "A uniplanar compact photonic-bandgap (UC-PBG) structure and its applications for microwave circuit," *IEEE Trans. Microwave Theory & Tech.*, vol. 47, no. 8, pp. 1509-1514, August 1999.

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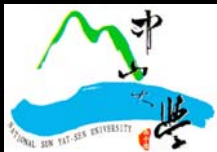


Introduction

- Although a design of the inductance-enhanced high impedance surface (HIS)[2] and the concept of cascading EBG structure with different stop-bands were proposed to achieve wider bandgap bandwidth, there are some drawbacks.

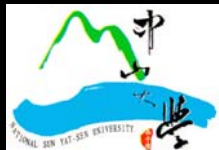
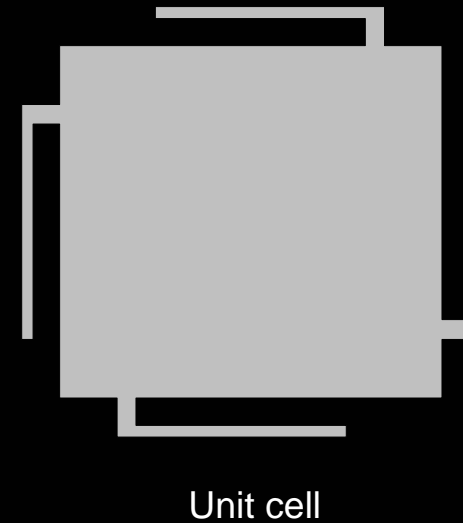
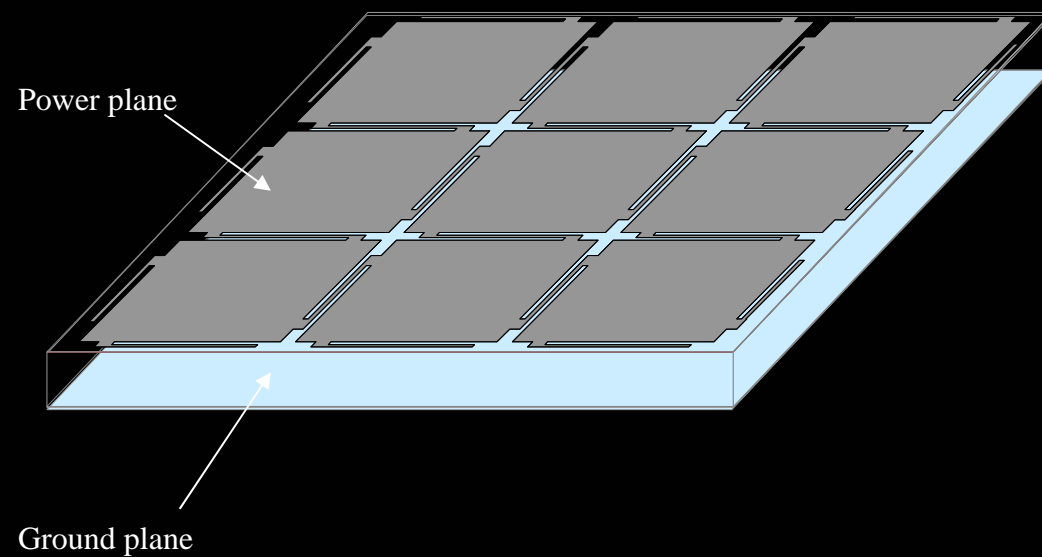


[2] T. Kamgaing, and O. M. Ramahi, "A novel power plane with integrated simultaneous switching noise mitigation capability using high impedance surface," *IEEE Microwave and Wireless Components Letters*, vol. 13 no. 1 pp. 21-23, January 2003



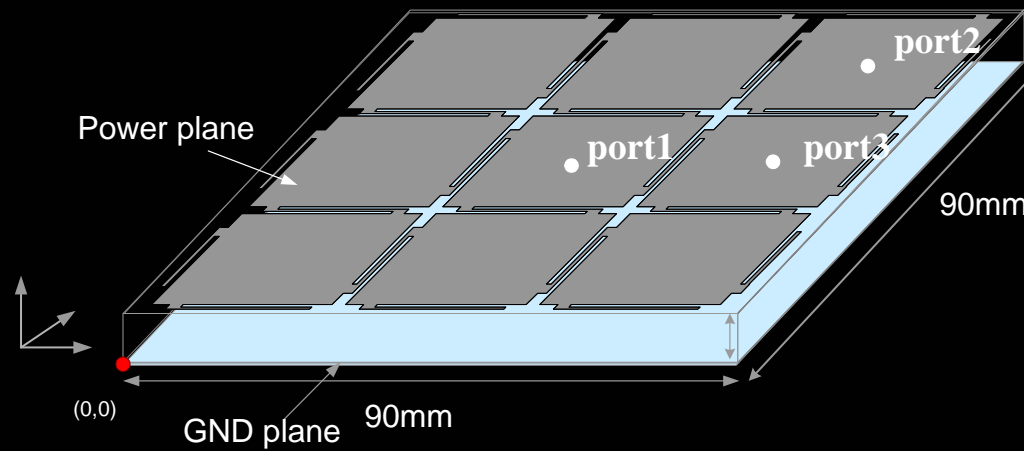
Introduction

- A Novel power plane designed with a coplanar EBG structure is proposed in this work with 4GHz stop-band covering from 600MHz to 4.6GHz.

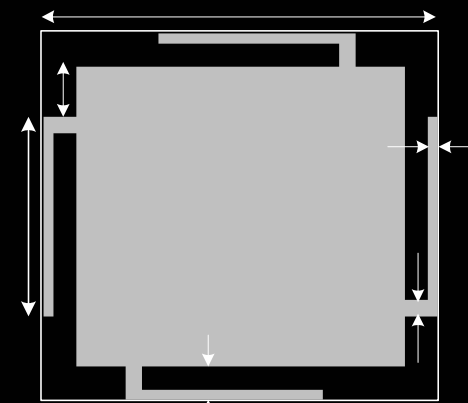


L-bridge EBG design concept

L-bridge EBG structure

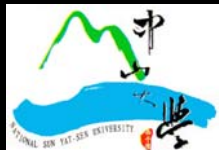


Nine cell L-bridge EBG

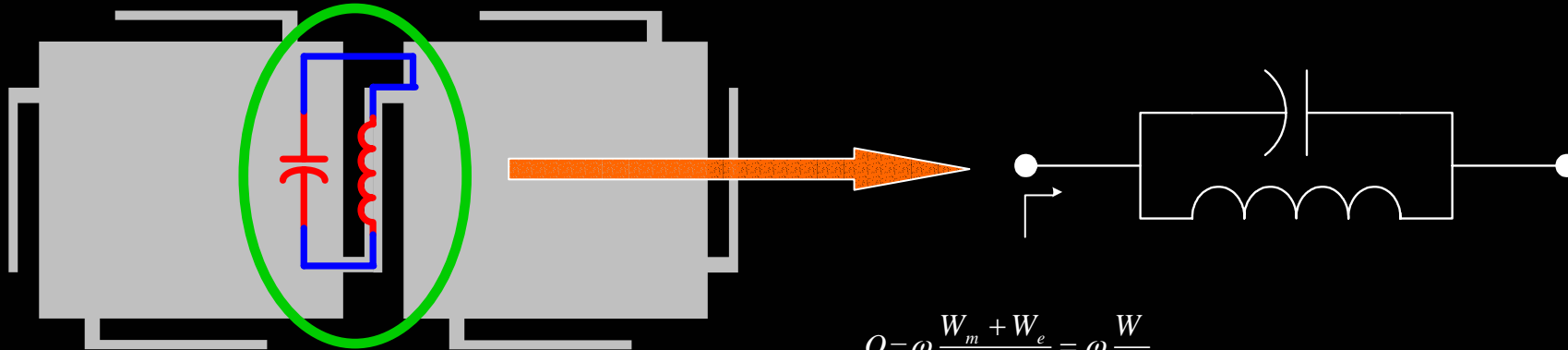


Unit cell

- $a = 30mm$
- $w = a / 4 = 7.5mm$
- $g1 = 0.1mm$
- $g2 = 0.2mm$
- $g3 = 0.65mm$
- $L = 15.2mm$



L- bridge EBG design concept



$$Z_{in} = \frac{1}{j\omega C} \parallel j\omega L = \frac{j\omega L}{1 - \omega^2 LC}$$

When $\omega = \omega_0 = \frac{1}{\sqrt{LC}}$ $Z_{in} \rightarrow \infty$

$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

$$Q = \omega \frac{W_m + W_e}{P_L} = \omega \frac{W}{P_L}$$

W : total time average energy in a cavity resonator

W_e : the energy stored in the electric fields

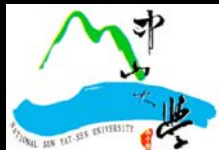
W_m : the energy stored in the magnetic fields

P_L : time-average power dissipated in the cavity

at resonant frequency $W_e = W_m$

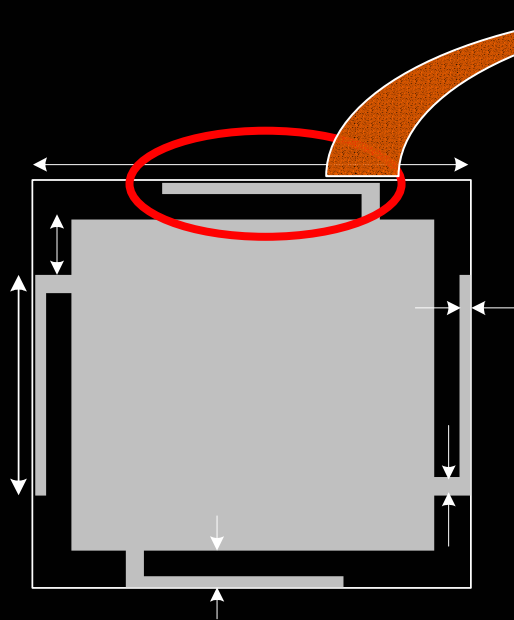
$$Q = \omega_0 \frac{2W_e}{P_L} = \omega_0 \frac{2 \cdot \frac{1}{4} C |V|^2}{\frac{1}{2} \frac{|V|^2}{R}} = \omega_0 RC = R \sqrt{\frac{C}{L}}$$

$$BW = \frac{1}{Q} = \frac{1}{R} \sqrt{\frac{L}{C}} \propto \sqrt{\frac{L}{C}}$$



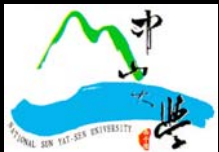
L- bridge EBG design concept

Design concept : $f_0 \propto \frac{1}{2\pi\sqrt{LC}}$ $BW \propto \sqrt{\frac{L}{C}}$



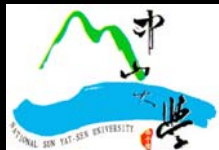
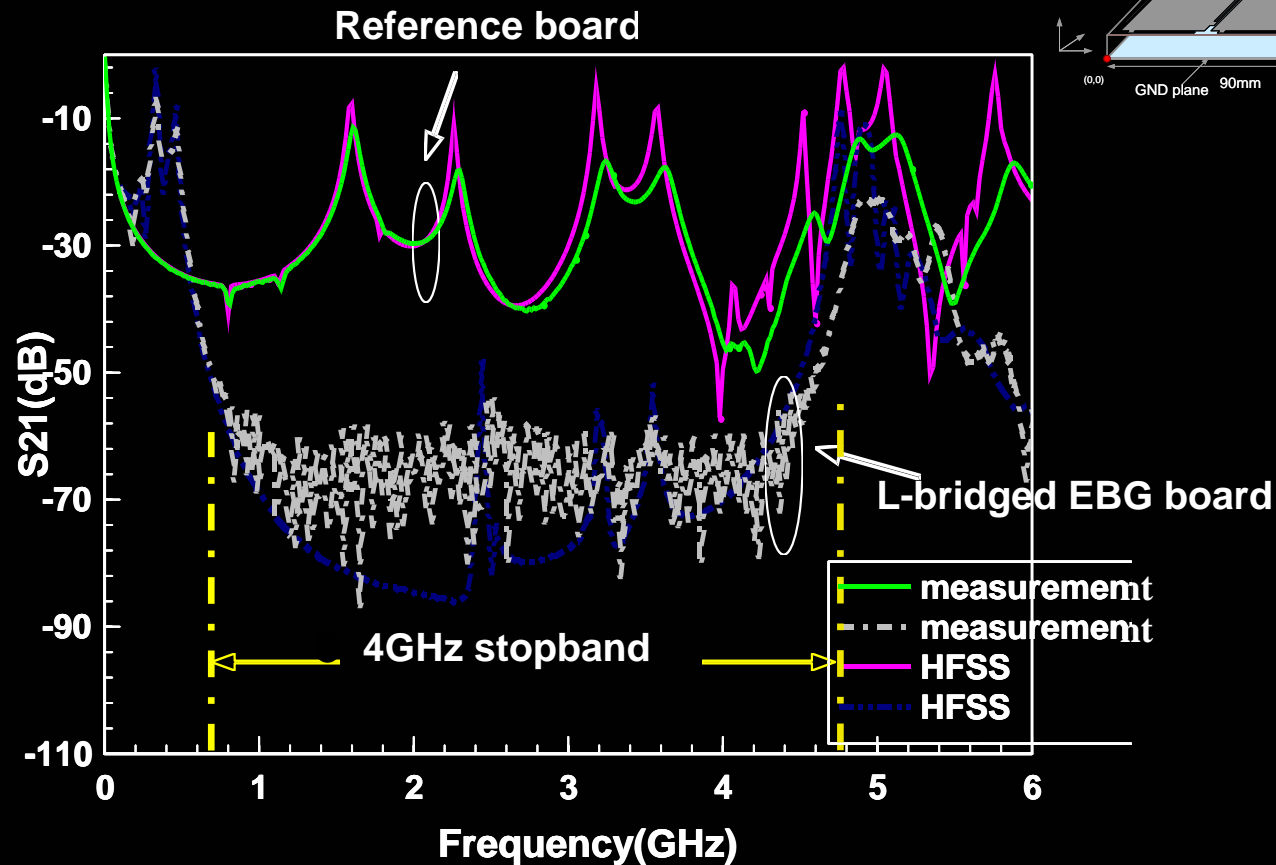
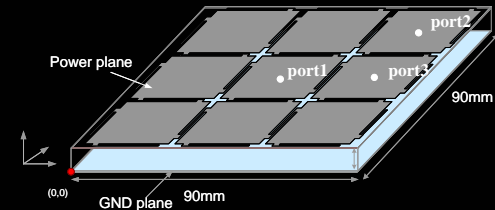
Unit cell

- Increase the effective inductance between adjacent cells and thus increase the stop-band bandwidth
- Ease the damage of the imperfect power plane to the signal quality.



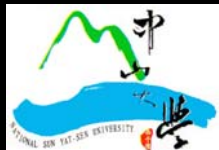
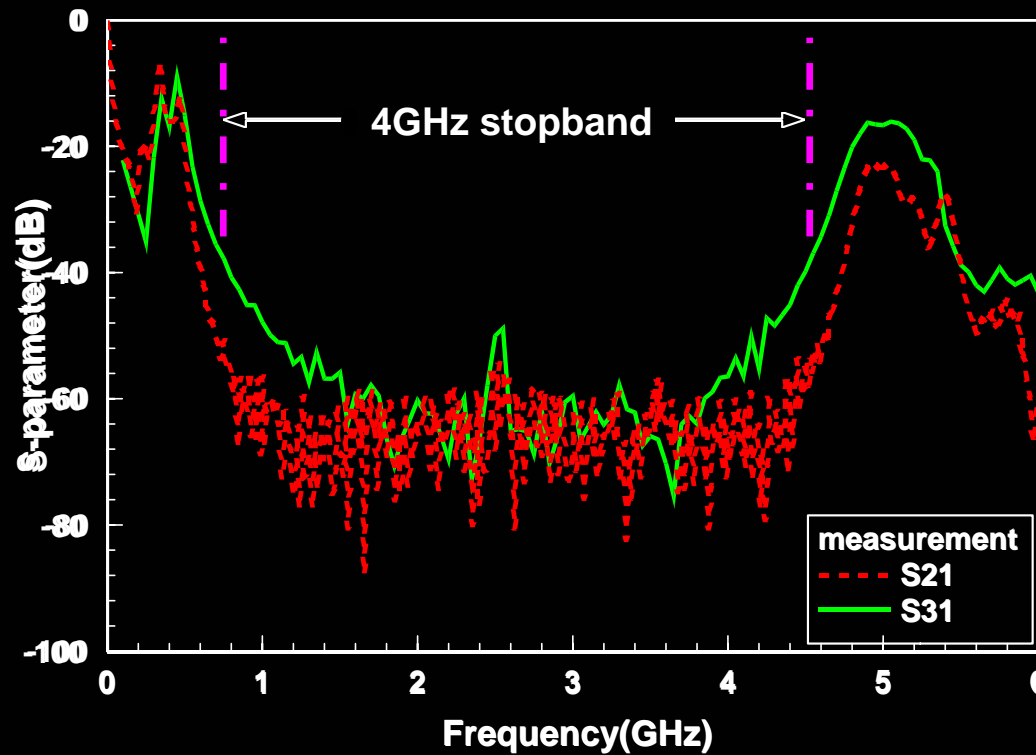
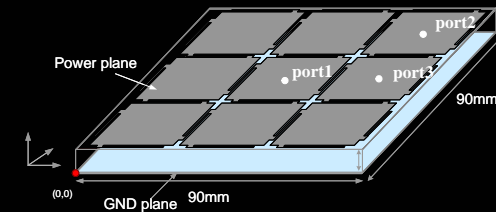
The simulation and measurement result

Broadband GBN suppression

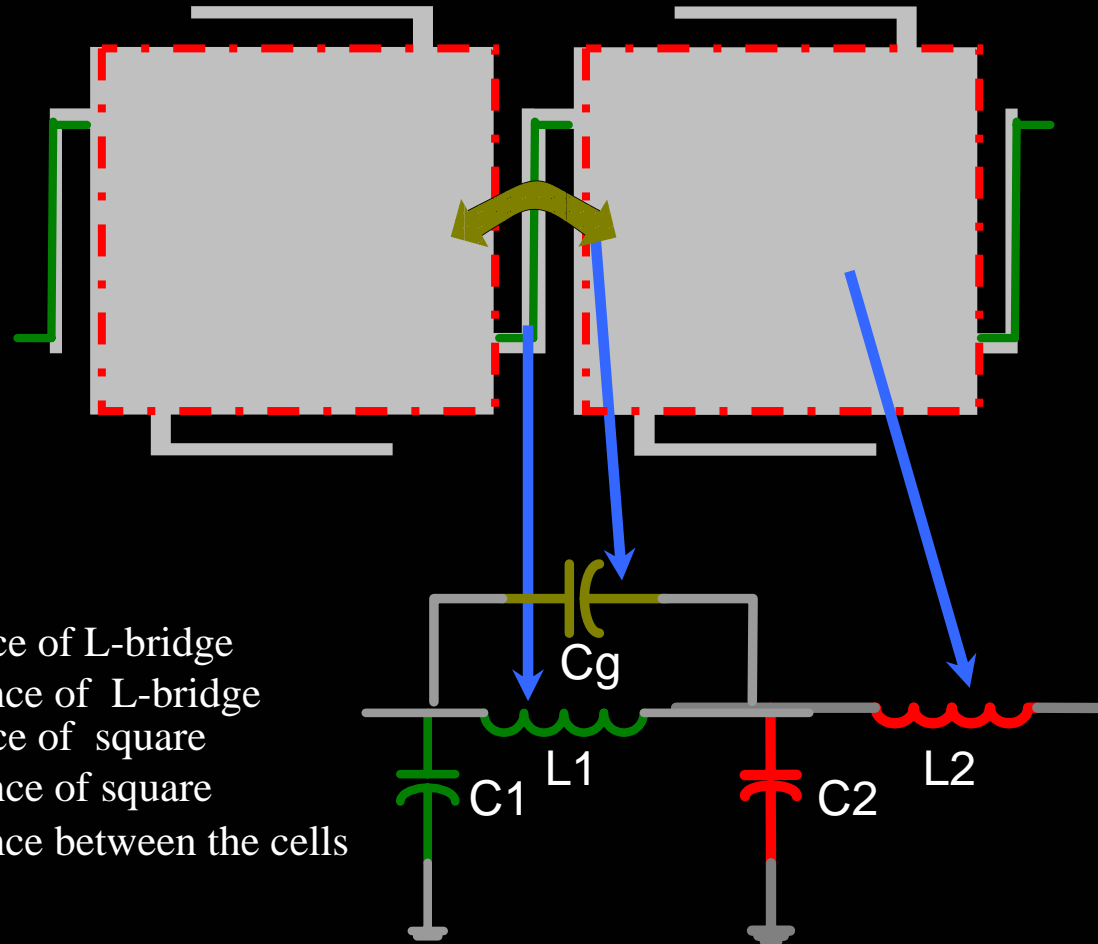


The simulation and measurement result

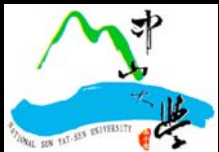
Broadband GBN suppression



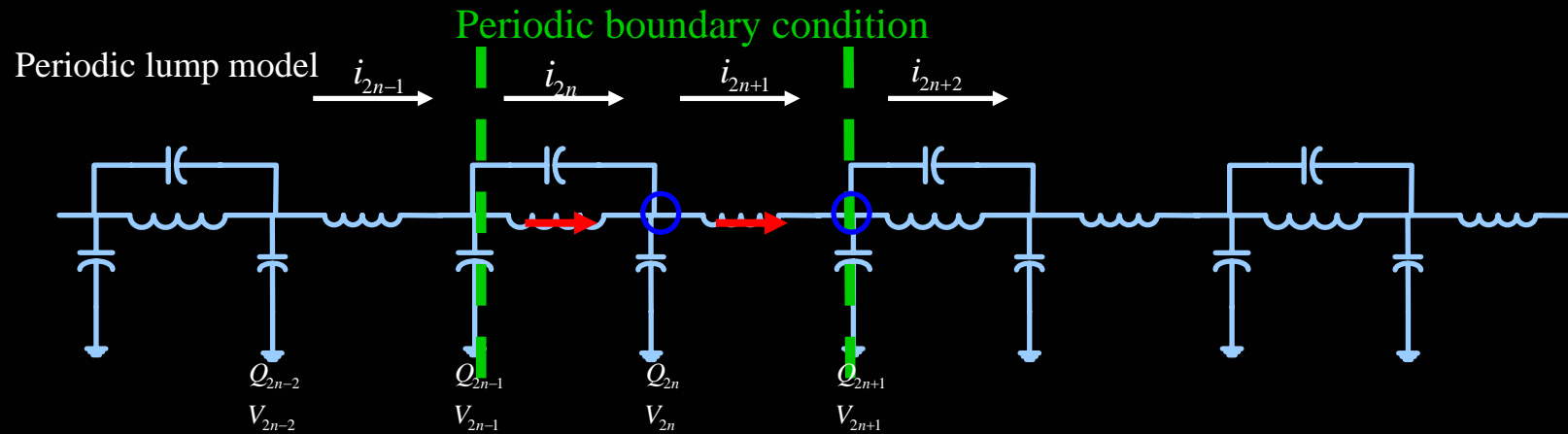
Stopband Model for the EBG Power Planes



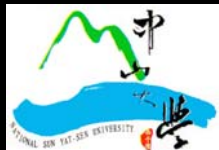
- $L1$: the inductance of L-bridge
- $C1$: the capacitance of L-bridge
- $L2$: the inductance of square
- $C2$: the capacitance of square
- Cg : the capacitance between the cells



Stopband Model for the EBG Power Planes



$$\left\{ \begin{array}{l} L_1 \frac{d}{dt} \frac{i_{2n}}{1 - \omega^2 L_1 C_g} = V_{2n-1} - V_{2n} = \frac{Q_{2n-1}}{C_1} - \frac{Q_{2n}}{C_2} \\ L_2 \frac{di_{2n+1}}{dt} = V_{2n} - V_{2n+1} = \frac{Q_{2n}}{C_2} - \frac{Q_{2n+1}}{C_1} \end{array} \right. \left\{ \begin{array}{l} i_{2n} - i_{2n+1} = \frac{dQ_{2n}}{dt} \\ i_{2n+1} - i_{2n+2} = \frac{dQ_{2n+1}}{dt} \end{array} \right.$$

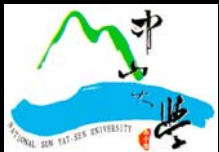


Stopband Model for the EBG Power Planes

$$\left\{ \begin{array}{l} L_1 \frac{d^2 \frac{i_{2n}}{1 - \omega^2 L_1 C_g}}{dt^2} = \frac{i_{2n-1} - i_{2n}}{C_1} - \frac{i_{2n} - i_{2n+1}}{C_2} \\ L_2 \frac{d^2 i_{2n+1}}{dt^2} = \frac{i_{2n} - i_{2n+1}}{C_2} - \frac{i_{2n+1} - i_{2n+2}}{C_1} \end{array} \right. \text{assume} \left\{ \begin{array}{l} i_{2n} = A_2 e^{i(\omega t - 2nk_1)} \\ i_{2n+1} = A_1 e^{i(\omega t - (2n+1)k_1)} \end{array} \right.$$



$$\left\{ \begin{array}{l} \left(\frac{-L_1 \omega^2}{1 - \omega^2 L_1 C_g} + \frac{1}{C_1} + \frac{1}{C_2} \right) A_2 + \left(\frac{1}{C_2} e^{-ik_1} - \frac{1}{C_1} e^{ik_1} \right) A_1 = 0 \\ \left(-L_2 \omega^2 + \frac{1}{C_1} + \frac{1}{C_2} \right) A_1 + \left(\frac{1}{C_2} e^{ik_1} - \frac{1}{C_1} e^{-ik_1} \right) A_2 = 0 \end{array} \right.$$

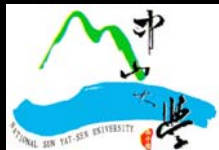


Stopband Model for the EBG Power Planes

$$\left\{ \begin{array}{l} \left(\frac{-L_1 \omega^2}{1 - \omega^2 L_1 C_g} + \frac{1}{C_1} + \frac{1}{C_2} \right) \left(-L_2 \omega^2 + \frac{1}{C_1} + \frac{1}{C_2} \right) \\ - \left(\frac{1}{C_2} e^{-ik_1} + \frac{1}{C_1} e^{ik_1} \right) \left(\frac{1}{C_2} e^{ik_1} + \frac{1}{C_1} e^{-ik_1} \right) = 0 \end{array} \right.$$

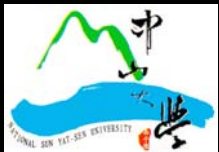
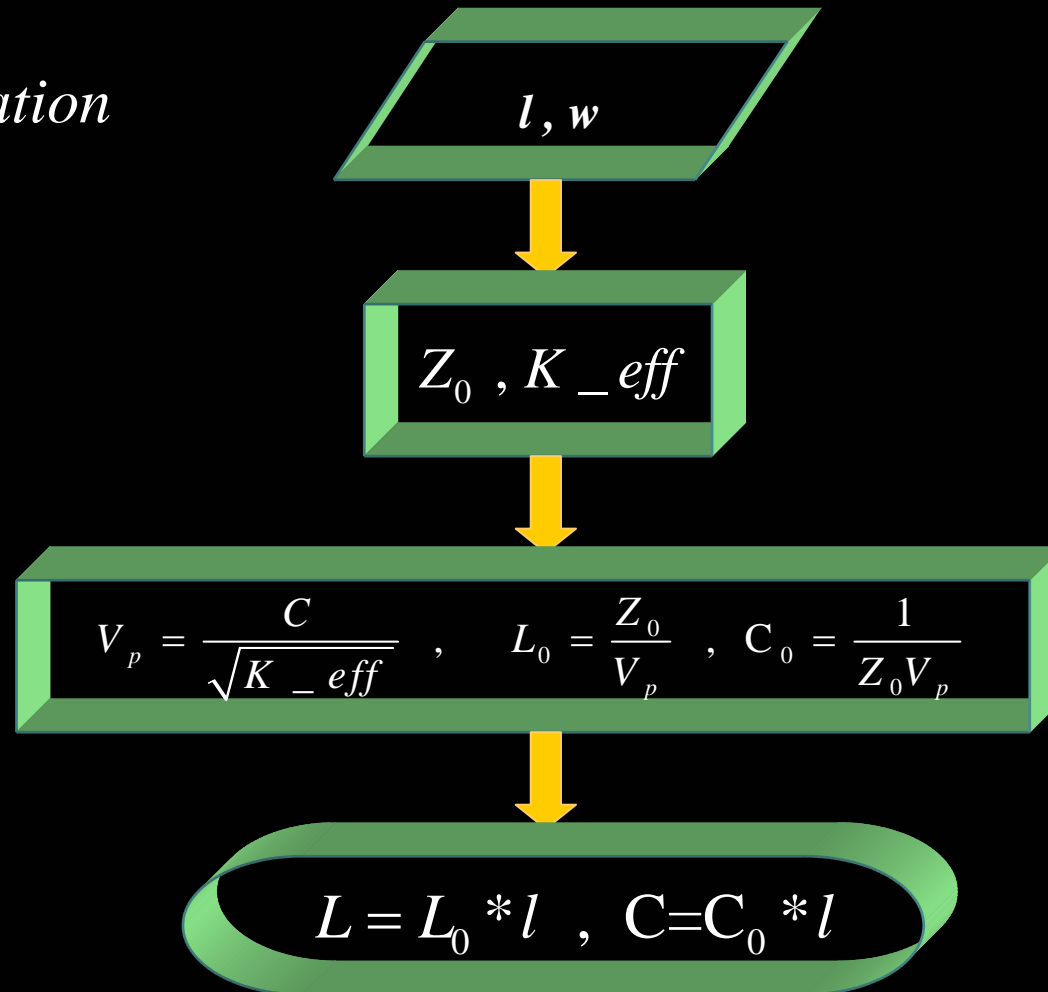


$$\omega = f(k)$$



Stopband Model for the EBG Power Planes

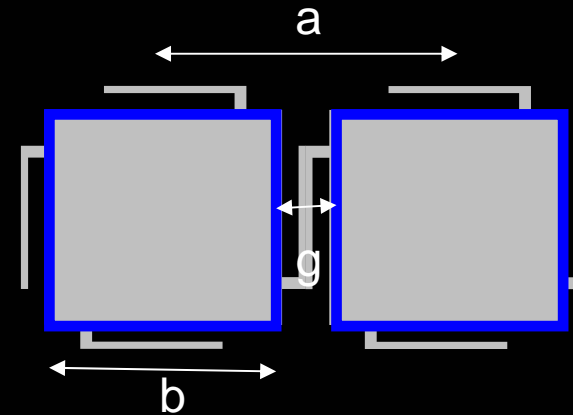
L_C estimation



Stopband Model for the EBG Power Planes

C_g estimation

$$C_g = b \frac{\epsilon_0 (1 + \epsilon_r)}{\pi} \text{Cosh}^{-1} \frac{a}{g}$$



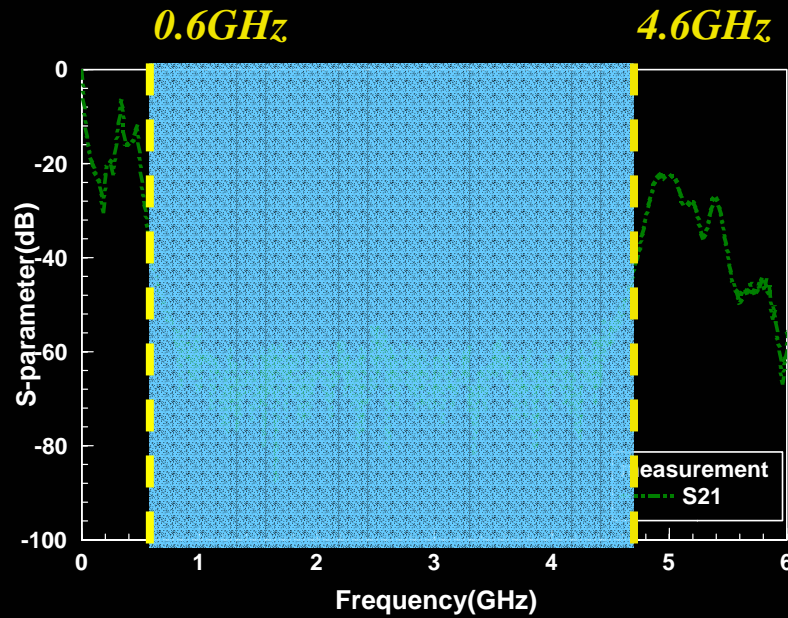
$$a = 30\text{mm} \quad b = 28.5\text{mm} \quad g = 1.5\text{mm}$$

$$\epsilon_r = 4.4$$

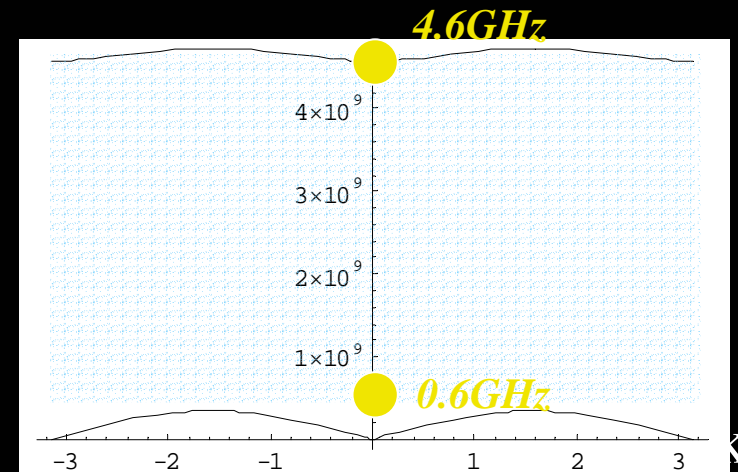
➔ $C_g = 1.59757 \text{ pF}$



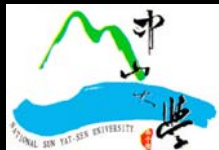
Stopband Model for the EBG Power Planes



measurement

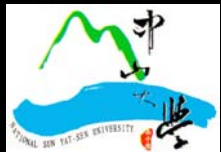
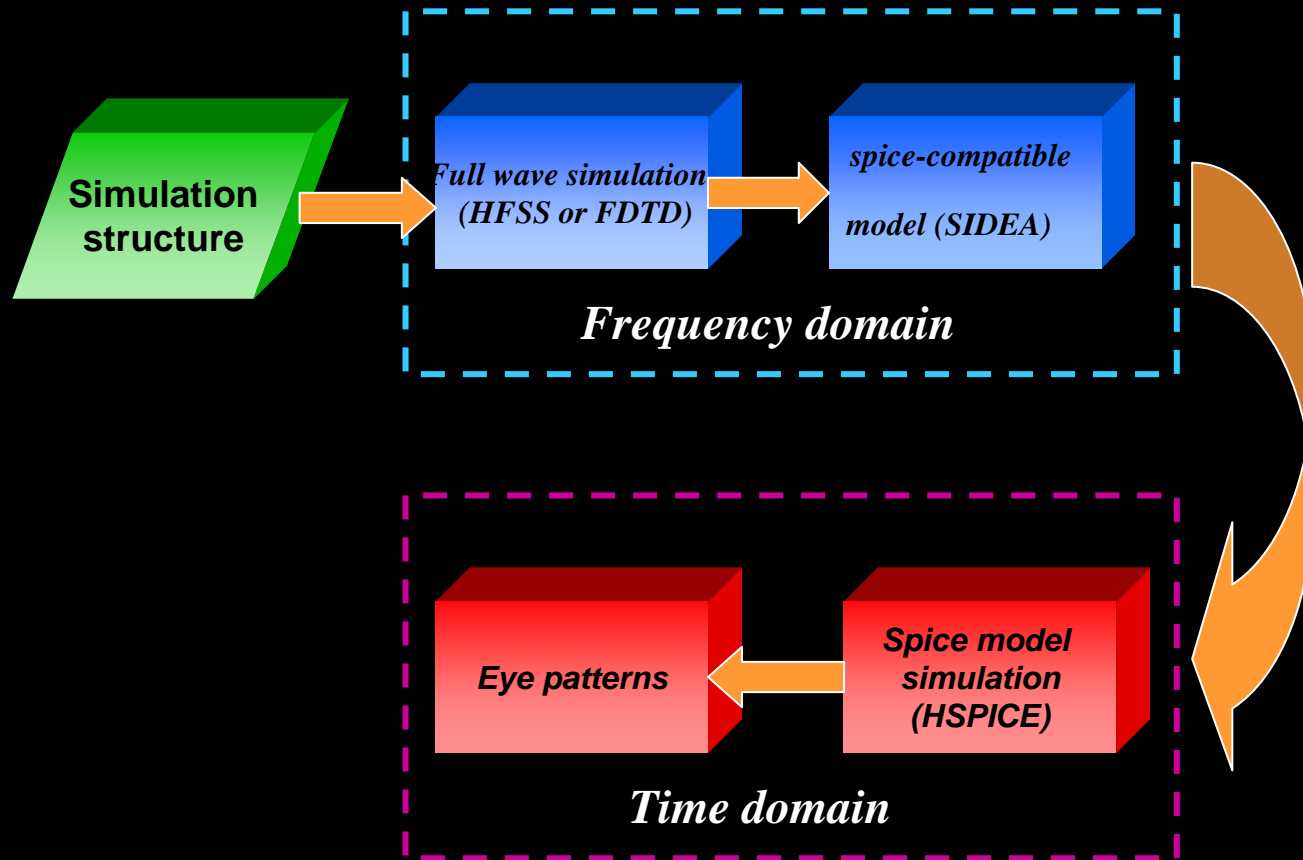


Dispersion diagram



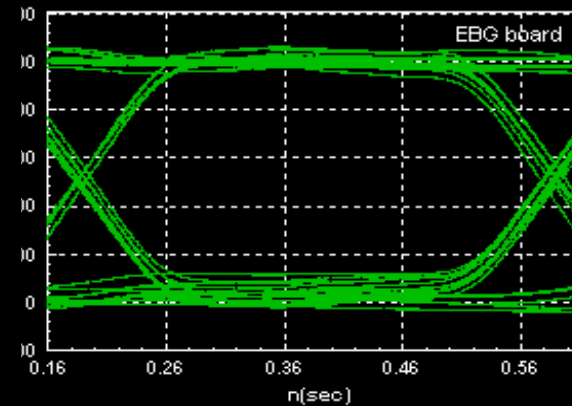
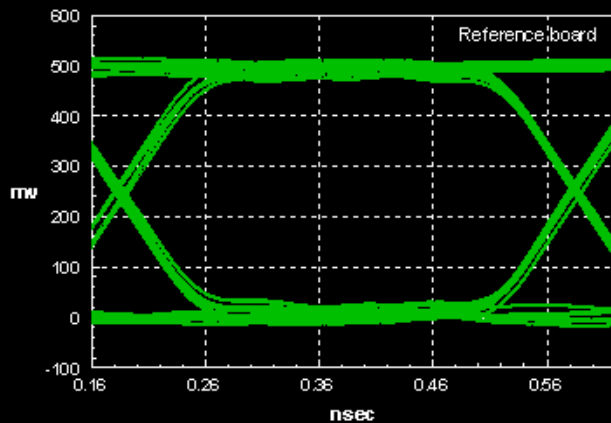
Signal integrity for L-bridge EBG power plane

Flowchart

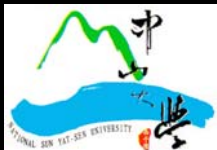


Signal integrity for L-bridge EBG power plane

Source : 2^7-1 PRBS, NRZ, coded at 2.5Gb/s, bit sequence swing is 500mv, rising/falling time 120ps

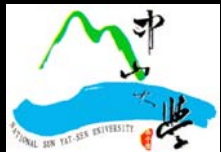
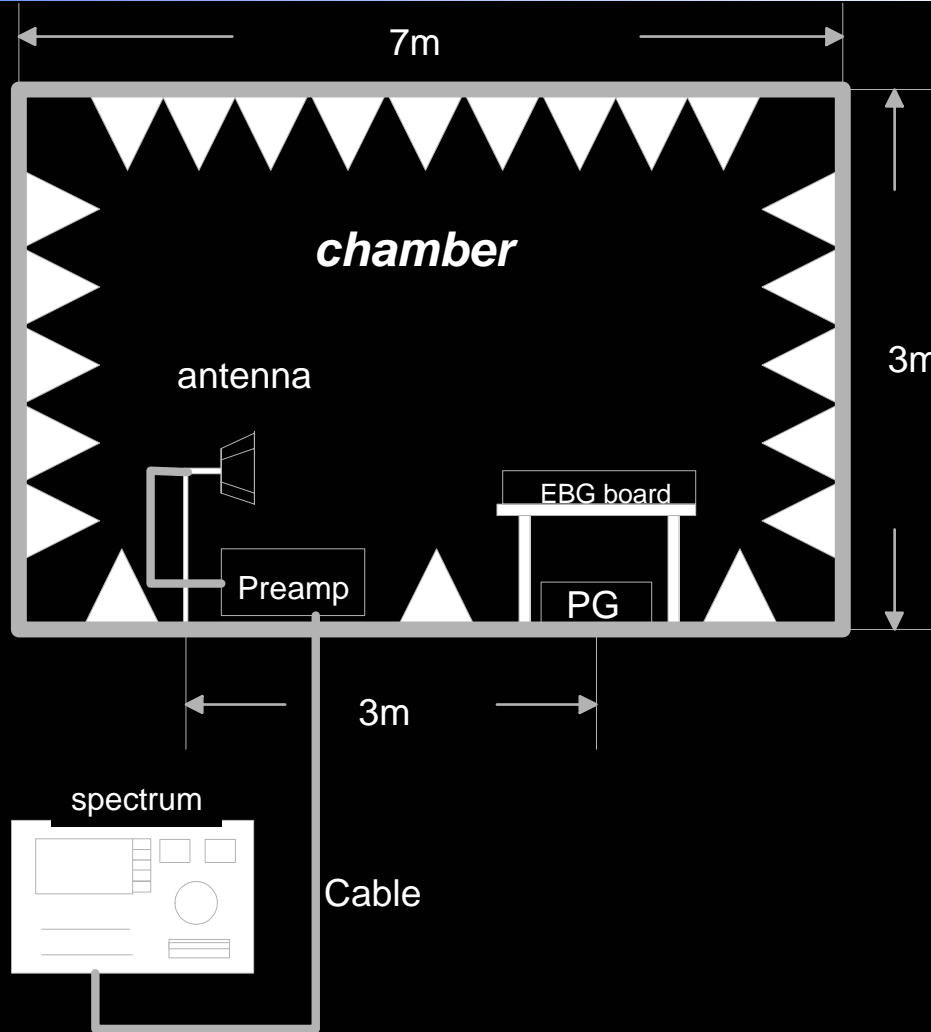


	Maximum eye opening (mv)	Maximum eye width (ps)
NO EBG	448.8	390
EBG	412.3	382



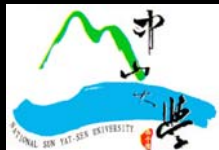
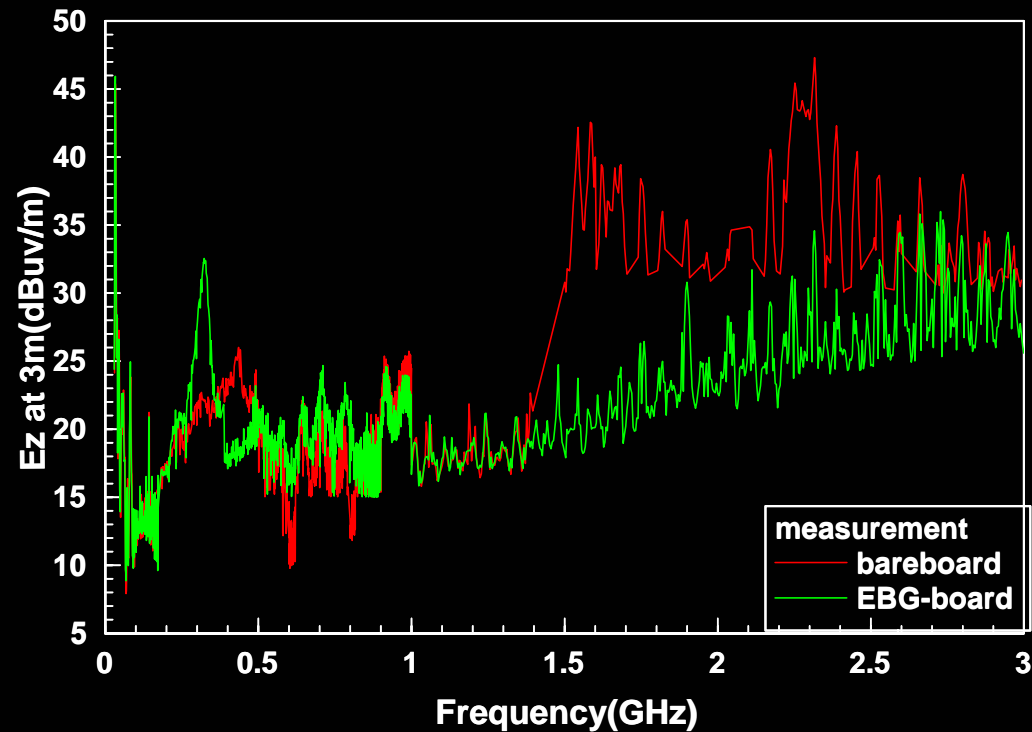
EMI for L-bridge EBG power plane

Set up



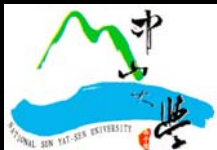
EMI for L-bridge EBG power plane

Measurement result



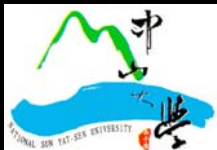
Conclusion

- A novel L-bridged EBG power plane is proposed in this paper with super-wideband suppression of the GBN from 600MHz to 4.6GHz.
- Compared with previous designs, this novel structure provides three advantage.
 - a. The L-bridged power plane broadens the stop band bandwidth to 4GHz and can cover to the low frequency range of 600MHz.
 - b. The signal quality is still kept acceptably good for the signal referring to the perforated EBG power plane
 - c. It is cost effective because only two layer metals are needed to design this novel power/ground plane structure.
- The excellent performance of the low-period PBG power planes is investigated both by measurement and simulation. Good agreement is seen.

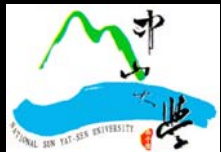


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Thanks for Your Attention



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